INTEGRATED CIRCUITS

DATA SHEET

SA56613-XX 5 V, 150 mA LDO and independent delayed RESET

Product data Supersedes data of 2002 Mar 25





5 V, 150 mA LDO and independent delayed RESET

SA56613-XX

DESCRIPTION

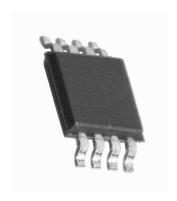
The SA56613-XX has a precise fixed 5 V output with a typical accuracy of $\pm 2\%$ to provide very low dropout and low noise in CD-ROM drives, battery-operated systems, and portable computers applications. This regulator consists of an internal voltage reference, an error amplifier, a driver with current limiter and a thermal shutdown.

An Active-LOW RESET is asserted when the regulator output voltage (V_{OUT}) falls below the reset detection voltage threshold. The SA56613-XX is available with fixed detection threshold voltages of 4.2, 4.5, and 4.7 V. The RESET output remains low for 1 ms (typical) when a 10 nF capacitor is connected to C_D pin. The reset time delay can be adjusted by replacing capacitance values from C_D pin to Ground.

The device is available in the small SO8 package (SOP005).



- Very low dropout voltage: 250 mV typ. (I_{out} = 30 mA)
- High precision output voltage: ±2%
- Output current capacity: 150 mA
- \bullet Low Noise: 200 μV_{rmS} typ. @ 20 Hz to 80 kHz and for $I_{OUT} = 30$ mA
- Extremely good line regulation: 10 mV typical
- Very good load regulation: 40 mV typical
- Low temperature drift co-efficient to V_{OUT}: ±100 ppm/°C
- Internal current limit and thermal shut-down circuits
- Adjustment-free reset detection voltages: 4.2 V typ., 4.5 V typ. and 4.7 V typ.
- Delay time can be adjusted by external capacitor.
- Wide operating temperature range: −40 °C to +85 °C



APPLICATIONS

- TV and monitors
- Electronic notebooks, PDAs and Palmtop computers
- Cameras, VCRs and camcorders
- PCMCIA cards and CD-ROM drives
- Modems
- Battery-powered or hand-held instruments

SIMPLIFIED SYSTEM DIAGRAM

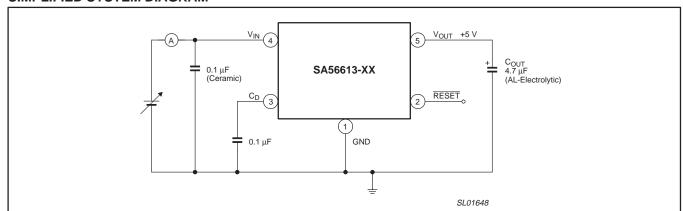


Figure 1. Simplified system diagram.

5 V, 150 mA LDO and independent delayed $\overline{\text{RESET}}$

SA56613-XX

ORDERING INFORMATION

TYPE NUMBER	PACKAGE	TEMPERATURE		
TIPE NOMBER	NAME	DESCRIPTION	VERSION	RANGE
SA56613- XX D	SO8	plastic small outline package; 8 leads; body width 3.9 mm SOP005		–40 to +85 °C

NOTE:

The device has 3 voltage options, indicated by the XX on the

^{&#}x27;Type number'.

XX	OUTPUT VOLTAGE (Typical)	RESET THRESHOLD (Typical)
42	5.0 V	4.2 V
45	5.0 V	4.5 V
47	5.0 V	4.7 V

PIN CONFIGURATION

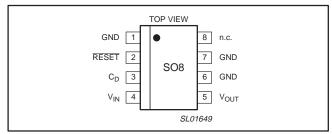


Figure 2. Pin configuration.

PIN DESCRIPTION

PIN	SYMBOL	DESCRIPTION
1	GND	Ground
2	RESET	Active-LOW reset signal output pin. The output remains LOW while V _{OUT} is below V _{SH} , the reset threshold, and for an external set time delay C _D pin after V _{OUT} rises above reset threshold.
3	C _D	Reset delay time capacitor pin. RESET pin output delay time can be set by the capacitance connected to the C_D pin. $t_{PLH} = 10^5 \times C$ where: $t_{PLH} = t_{PLH} = $
4	V _{IN}	Supply voltage input pin
5	V _{OUT}	Regulated output voltage pin
6	GND	Ground pin and heat sink
7	GND	Ground pin and heat sink
8	n.c.	not connected

MAXIMUM RATINGS

SYMBOL	PAR	AMETER	MIN.	MAX.	UNIT
V _{IN}	Input supply voltage		-0.3	+18	V
I _{OUT}	Output current		-	200	mA
T _{opr}	Operating ambient temperature		-40	+85	°C
T _{stg}	Storage temperature		-40	+150	°C
T _{j(max)}	Maximum junction temperature		-	+125	°C
P _D	Power dissipation (Note 1) Derate 6.5 mW/°C above T _{amb} 25 °C		-	650	mW

NOTE:

1. When mounted on a 55×20 mm paper phenol board.

5 V, 150 mA LDO and independent delayed $\overline{\text{RESET}}$

SA56613-XX

ELECTRICAL CHARACTERISTICS

 T_{amb} = 25 °C, Figure 13 "Test circuit", unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
I _{CCq1}	No-load input current 1	V _{IN} = 6 V; I _{OUT} = 0 mA	_	400	800	μА
I _{CCq2}	No-load input current 2	V _{IN} = 4 V; I _{OUT} = 0 mA	_	2.5	_	μΑ
Regulator	•					
V _{OUT}	Output voltage	V _{IN} = 6 V; I _{OUT} = 30 mA	4.90	5.00	5.10	V
V _{i/o(dif)}	Input/output differential voltage	V _{IN} = 4.8 V; I _{OUT} = 150 mA	-	0.25	0.5	V
ΔV_1	Line regulation	$V_{IN} = 6 \text{ V} \rightarrow 10 \text{ V}; I_{OUT} = 30 \text{ mA}$	-	10	30	mV
ΔV_2	Load regulation	$V_{IN} = 6 \text{ V}; I_{OUT} = 0 \text{ mA} \rightarrow 150 \text{ mA}$	-	40	80	mV
$\Delta V_{OUT}/\Delta T$	V _{OUT} temperature coefficient (Note 1)	$ T_{j} = -20 ^{\circ}\text{C} \rightarrow +85 ^{\circ}\text{C}; $ $ V_{IN} = 6 \text{V}; I_{OUT} = 30 \text{mA} $	-	100	-	ppm/°C
RR	Ripple rejection (Note 1)	$V_{IN} = 6 \text{ V; } f = 120 \text{ Hz;}$ $V_{RIPPLE} = 1 \text{ V}_{p-p}; I_{OUT} = 30 \text{ mA}$	50	60	-	dB
V _{n(o)}	Noise output voltage (Note 1)	$V_{IN} = 6 \text{ V; } f = 20 \sim 80 \text{ kHz;} $ $I_{OUT} = 30 \text{ mA}$	_	200	400	μV_{rms}
Reset	-	-				
Vs	Detection voltage	$V_{IN} = H \rightarrow L$ SA56613-42 SA56613-45 SA56613-47	4.03 4.31 4.51	4.20 4.50 4.70	4.37 4.69 4.89	V V V
ΔV _S /ΔT	V _S temperature coefficient (Note 1)	$T_j = -20 ^{\circ}\text{C} \rightarrow +85 ^{\circ}\text{C}$	-	100	-	ppm/°C
ΔV _S	Hysteresis voltage	$V_{IN} = H \rightarrow L \rightarrow H$	25	50	100	mV
V _{OL}	LOW-level output voltage	$V_{IN} = 3.9 \text{ V}; R_L = 4.7 \text{ k}\Omega$	-	100	200	mV
t _{PLH1}	Reset delay time	V_{IN} = 4 V \rightarrow 5 V; C_D = 0.1 μF	5	10	15	ms
t _{PHL}	'L' transmission delay time (Note 1)	V_{IN} = 5 V \rightarrow 4 V; C_D = 0.1 μF	-	30	90	μs
V _{OPL}	Threshold operating voltage	V _{OL} = 0.4 V	-	0.65	0.85	V

NOTE:

^{1.} This parameter is guaranteed by design.

5 V, 150 mA LDO and independent delayed RESET

SA56613-XX

TYPICAL PERFORMANCE CURVES (SA56613-42D)

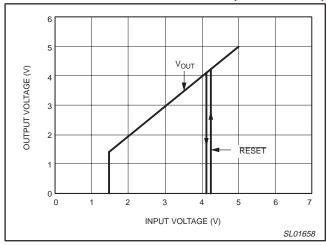
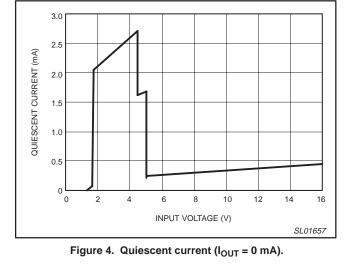


Figure 3. Detection voltage ($I_{OUT} = 0 \text{ mA}$).



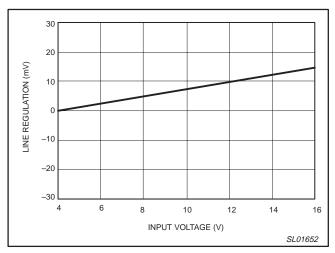


Figure 5. Line regulation.

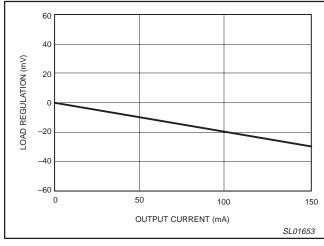


Figure 6. Load regulation.

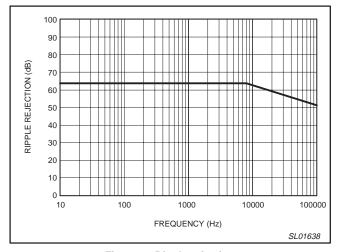


Figure 7. Ripple rejection.

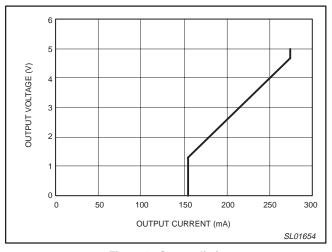


Figure 8. Current limit.

5 V, 150 mA LDO and independent delayed RESET

SA56613-XX

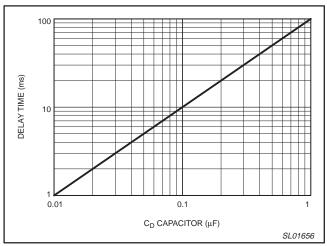


Figure 9. Reset delay time.

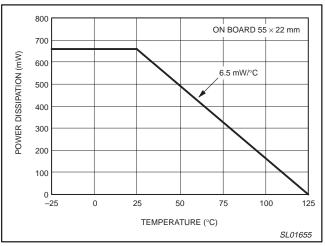


Figure 10. Power dissipation.

5 V, 150 mA LDO and independent delayed RESET

SA56613-XX

TIMING DIAGRAM

The Timing Diagram in Figure 11 depicts the operation of the device. Letters A-I on the Time axis indicates specific events.

- **A:** At "A", V_{OUT} abruptly begins to increase. Also the $\overline{\text{RESET}}$ voltage initially increases but abruptly decreases when V_{OUT} reaches the threshold operating level (typically 0.65 V) that activates the internal bias circuitry and $\overline{\text{RESET}}$ is asserted.
- **C:** At "C", V_{OUT} is above V_{SL} and the delay time, t_{PLH} elapses. At this instant, the device releases the hold on the \overline{RESET} . The reset output then goes HIGH. In a microprocessor based system these events release the reset from the microprocessor, allowing the microprocessor to function normally.
- **D-E:** At "D", V_{IN} falls below 5 V, causing V_{OUT} to follow. V_{OUT} continues to fall until the V_{SL} undervoltage detection threshold is reached at "E". This causes a reset signal to be generated (RESET goes LOW).

- **E-F:** Between "E" and "F", V_{OUT} continues to fall and then starts rising.
- **F:** At "F", V_{OUT} rises to the V_{SH} level. Once again, the device initiates the delay timer.
- **F-G:** V_{OUT} rises above V_{SH} and returns to normal 5 V output. At "G", the delay (t_{PLH}) times out and once again, then it releases the hold on the RESET and it goes HIGH.
- **G-H:** At "G", V_{OUT} is above the upper threshold and begins to fall, causing \overline{RESET} to follow it. As long as V_{OUT} remains above the V_{SL} , no reset signal will be generated.
- **H:** At event "H", V_{OUT} falls until the V_{SL} undervoltage detection threshold is reached. At this level, a RESET signal is generated and RESET goes LOW.
- I: At event "I", V_{OUT} has decreased until normal internal circuit bias is unable to maintain a RESET. As a result, V_{CC} may rise to less than 0.65 V. As V_{CC} decreases further, the V_{OUT} reset also decreases to zero.

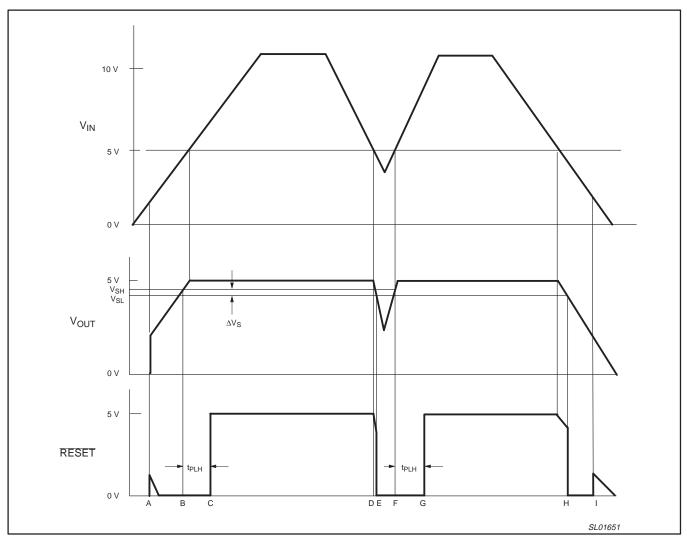


Figure 11. Timing diagram.

5 V, 150 mA LDO and independent delayed RESET

SA56613-XX

APPLICATION INFORMATION

Input capacitor

An input capacitor of \geq 1 μ F is required to eliminate the AC coupling noise. This capacitor must be located as close as possible to V_{IN} or GND pin (not more than 1 cm) and returned to a clean analog ground. Any good quality ceramic, tantalum or film capacitor will work.

Output capacitor

Phase compensation is made for securing stable operation even if the load current varies. For this reason, an output capacitor with good frequency characteristics is needed. Set it as close to the circuit as possible and make the wiring as short as possible.

The value of the output capacitance has to be at least 47 μ F connected from V_{OUT} to GND. When operating from sources other than batteries, supply-noise rejection and transient response can be improved by increasing the value of the input and output capacitors and employing passive filtering techniques.

RESET output

The SA56613-XX has an Active-LOW RESET output. When V_{OUT} of the regulator rises above V_{SH} , the upper detection threshold voltage, the reset delay time is initiated. After the programmed delay time elapses, RESET is released and goes HIGH. The time delay can be set typically from 1 ms to 100 ms by connecting a time delay capacitor from C_D (pin 3) to ground (see Figure 9: RESET Delay time versus C_D , Delay Capacitor).

The RESET delay time (t_{PLH}) can be approximated by the following equation:

$$t_{PIH} = 10^5 \times C$$
 Eqn. (1)

(Time is expressed in seconds, Capacitance in Farads.)

For example, for a delay capacitor, C_D of 0.1 μF (100 nF), t_{PLH} is approximately 10 ms.

When the regulator output Voltage falls to or below V_{SL} , the lower detection threshold voltage, the RESET output is asserted and it goes to an Active-LOW state. This "LOW" transmission delay time is typically 30 μ s with C_D at 100 nF.

Reset hysteresis voltage

The reset hysteresis voltage, $\Delta V_{\mbox{\scriptsize S}}$ is defined in the following equation:

$$\Delta V_S = V_{SH} - V_{SI}$$
 Eqn. (2)

Hysteresis voltage is typically 50 mV. This small level of hysteresis ensures that the reset will not dither when the regulator V_{OUT} is noisy.

PCB layout

The component placement around the LDO should be done carefully to achieve good dynamic line and load response. The input and noise capacitors should be kept close to the LDO.

The rise in junction temperature depends on how efficiently the heat is carried away from the junction to ambient. The junction to lead thermal impedance is a characteristic of the package and fixed. The thermal impedance between lead to ambient can be reduced by increasing the copper area on the PCB. Increase the input, output and ground trace area to reduce the junction-to-ambient impedance.

Power dissipation

The SA56613-XX maximum power dissipation depends on the thermal resistance from the die junction to the ambient air. The maximum power dissipation shown in Figure 10 is 650 mW at ambient temperature of 25 $^{\circ}$ C. It is derated at 6.5 mW/ $^{\circ}$ C above 25 $^{\circ}$ C.

Power dissipation of the device is $P_D = I_{OUT} (V_{IN} - V_{OUT})$. The maximum power dissipation is:

$$P_{\text{max}} = \frac{(T_j - T_{\text{amb}})}{(\theta_{JA})}$$
 Eqn. (3)

where

 $\theta_{JA} = \theta_{JB} + \theta_{BA}$, the junction-to-ambient thermal resistance, θ_{JA} calculated from Figure 10 is 154 °C/W;

 θ_{JB} is the thermal resistance from the die junction to PCB material and copper traces;

 θ_{BA} is the thermal resistance from the PCB material and copper traces to the surrounding air.

The GND pin provides an electrical connection to ground and a path for heat transfer from the device to the PCB and to the surrounding air. To maximize heat transfer, connect the GND pin to a large ground pad or ground plane.

The following example determines the maximum I_{OUT} at T_{amb} = 25 $^{\circ}C$ for V_{IN} = 1 V.

$$I_{OUT} = \frac{P_D}{(V_{IN} - V_{OUT})} = \frac{650 \text{ mW}}{(12 - 5)} = 92.8 \text{mA}$$
 Eqn. (4)

The maximum output current of the SA56613-XX is reduced as the input voltage, V_{IN} and the ambient temperature, T_{amb} increase.

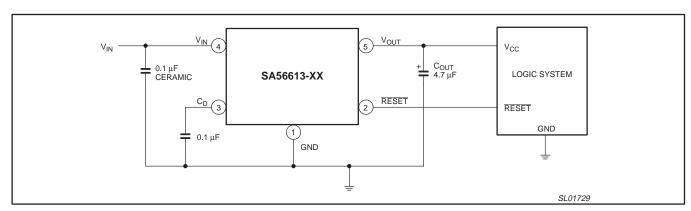


Figure 12. Typical application circuit.

5 V, 150 mA LDO and independent delayed RESET

SA56613-XX

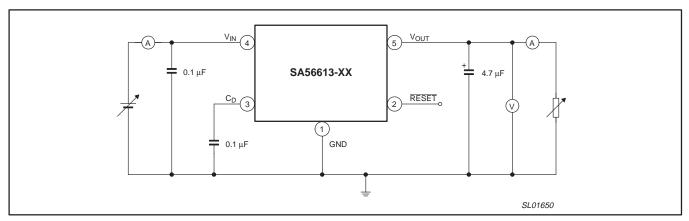


Figure 13. Test circuit.

PACKING METHOD

The SA56613-XX is packed in reels, as shown in Figure 14.

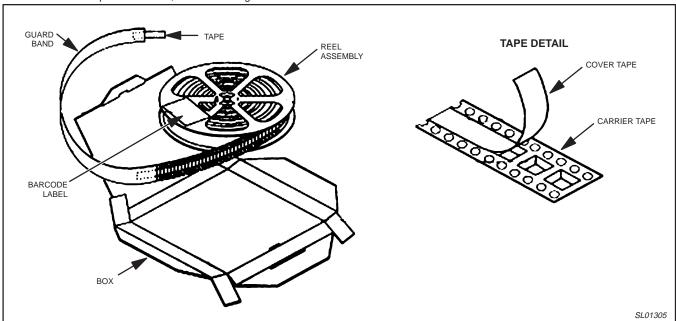


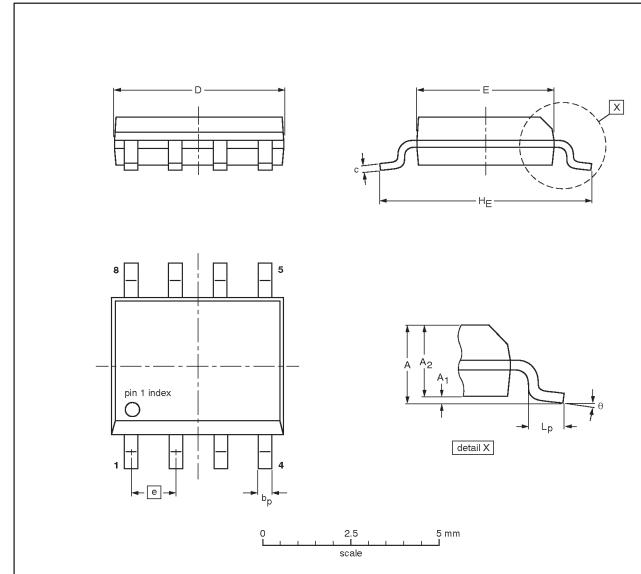
Figure 14. Tape and reel packing method.

5 V, 150 mA LDO and independent delayed RESET

SA56613-XX

SO8: plastic small outline package; 8 leads; body width 3.9 mm

SOP005



DIMENSIONS (inch dimensions are derived from the original mm dimensions)

UNIT	A max.	A ₁	A ₂	bp	С	D ⁽¹⁾	E ⁽²⁾	е	HE	Lp	θ
mm	1.73	0.25 0.10	1.45 1.25	0.51 0.33	0.25 0.19	4.95 4.80	4.0 3.8	1.27	6.2 5.8	1.27 0.38	8°
inches	0.068	0.010 0.004	0.057 0.049	0.013 0.020	0.0100 0.0075	0.20 0.19	0.16 0.15	0.050	0.244 0.228	0.050 0.015	0°

Notes

- 1. Plastic or metal protrusions of 0.15 mm maximum per side are not included.
- 2. Plastic or metal protrusions of 0.25 mm maximum per side are not included.

OUTLINE		REFER	EUROPEAN	ISSUE DATE			
VERSION	IEC	JEDEC	JEITA		PROJECTION	ISSUE DATE	
SOP005	076E03	MS-012				03-10-07	

5 V, 150 mA LDO and independent delayed RESET

SA56613-XX

REVISION HISTORY

Rev	Date	Description
_1	20031030	Product data (9397 750 12214). ECN 853-2331 30323 of 09 September 2003. Supersedes data of 2002 Mar 25 (9397 750 10039).
		Modifications:
		Change package outline version to SOP005 in Ordering information table and Package outline sections.
_1	20020325	Product data (9397 750 10039). ECN 853-2331 27919 of 25 March 2002.

Data sheet status

Level	Data sheet status ^[1]	Product status ^[2] [3]	Definitions
I	Objective data	Development	This data sheet contains data from the objective specification for product development. Philips Semiconductors reserves the right to change the specification in any manner without notice.
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^[1] Please consult the most recently issued data sheet before initiating or completing a design.

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